Magnetic Field Distribution of an Elliptical Permanent Magnet

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Abstract—The magnetic field distribution of an axially magnetised cylinder with an elliptical profile is analytically modelled and analysed in this paper. An accurate and fast-computed semi-analytical model is developed, based on the charge model and geometrical analysis, to compute three components of the magnetic field generated by this elliptical cylinder in three-dimensional space. The accuracy of the model is verified using Finite Element Analysis. The analytical expressions are efficient for calculating the implementation of the magnetic field, taking less than one millisecond to execute on a modern PC. Using the fast-computed analytical model, the distribution of the magnetic field of an axially magnetised cylinder with different elliptical profiles is studied and compared with that of a circular cylinder. The variations in magnetic field strength of axial, azimuthal and radial components can be used in novel sensing applications. The derived analytical model can be extended to calculate the magnetic field of arc-shaped elliptical and circular cylinders with axial magnetization, which can be used in Halbach arrangements.

1. INTRODUCTION

Understanding magnetic field distribution of a permanent magnet plays a key role in developing magnetic sensing devices [1] and medical appliances (i.e., magnetic resonance imaging) [2]. In other applications such as magnetic suspensions, levitated magnetic disks, magnetic gears, and magnetic couplings, the forces and torques generated by permanent magnets are of great interest [3]. These forces (\vec{F}) and torques (\vec{T}) can be computed using the charge model, involving the magnetic field [4] as:

$$\overrightarrow{F} = \iiint_{v} \overrightarrow{B}_{ext} \sigma_{mv} dv + \iint_{s} \overrightarrow{B}_{ext} \sigma_{ms} ds$$

and
$$\overrightarrow{T} = \iiint_{v} \left(\overrightarrow{r} \times \overrightarrow{B}_{ext} \right) \sigma_{mv} dv + \iint_{s} \left(\overrightarrow{r} \times \overrightarrow{B}_{ext} \right) \sigma_{ms} ds$$

where \overrightarrow{B}_{ext} is the external magnetic field; σ_{mv} is the volume charge of a given magnet with a magnetization vector \overrightarrow{J} ; $\sigma_{mv} = -\overrightarrow{\nabla} \cdot \overrightarrow{J}$; σ_{ms} is the surface charge of the given magnet; $\sigma_{ms} = \overrightarrow{J} \cdot \overrightarrow{n}$; \overrightarrow{n} is the normal vector to the impacted surface of the given magnet; \overrightarrow{r} is the distance between the magnetic field source and the given magnet; and V and S denote the volume and surface area of the given magnet, respectively. A fast-computed model of the magnetic field source generated by a permanent magnet is important to facilitate the design, optimization, and dynamical modelling of devices and systems using permanent magnets.

Although magnetic field solutions for generalized geometries are available using Finite Element Analysis (FEA), and this method is inefficient for 3D problems, with a substantial tradeoff between computation time and accuracy. To address this issue, magnetic field solutions of permanent magnets

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with specific shapes and configurations have been analytically modelled and studied [3–15]. Ravaud et al. [4] developed analytical models to calculate the magnetic field of ring-shaped permanent magnets with axial and radial magnetization. Xu et at. [5] studied the strong magnetic field of a permanent magnet array used in refrigeration. Ravaud et al. [6] derived the exact analytical model to compute the magnetic field produced by a radially magnetized tile permanent magnet. Jian and Chan [7] analytically calculated the magnetic field distribution of magnetic gears. Yonnet et al. [8] derived an analytical model to calculate permanent magnet couplings. McClurg [9] studied the magnetic field distribution of a sphere and ellipsoid. Based on a Coulombian approach, Rakotoarison et al. [10] modelled the scalar potential and magnetic field of an arc-shaped permanent magnet with radial magnetization. Nguyen and Lu [11] derived an analytical model to calculate the magnetic field of a circular cylinder with diametrical magnetization. Ravaud and Lemarquand [12] analysed the magnetic field of a parallelepipedic magnet. Nguyen and Lu [13] developed a fast-computed model to study the magnetic field distribution of an elliptical cylinder with diametrical magnetization. However, the distribution of the magnetic field of an elliptical cylinder with axial magnetization has not been analytically expressed and studied in the literature.

Inspired by the fact that elliptical profiles have been widely found in physics, astronomy, and engineering [13]. For example, the orbit of each planet in the solar system is approximately an ellipse [16]. Elliptical gears have been found to acquire great efficiency in motion and power transmission solutions [17–19]; hence, the magnetic field distribution and application of a permanent magnet with elliptical profiles could be of great interest [9]. Moreover, a fast-computed model to calculate the magnetic field could facilitate the design process of an elliptical permanent magnet used for surface magnetic resonance imaging [20], a novel magnetic coupler [21], the design of permanent magnetic gears [7], and couplings [8] with elliptical profiles as well as a non-contact cam mechanism [22] using an axially magnetised driver with an elliptical profile. Furthermore, for educational purposes, this research would assist students in understanding the magnetic field distribution of the elliptical cylinder permanent magnet as an example beyond the common permanent magnets in cuboid, ring, and circular cylindrical shapes, which are commonly found in textbooks [3]. In addition, having in hand a model of the magnetic field of an elliptical cylinder with axial and diametrical magnetizations, the distribution of the magnetic field of an elliptical cylinder with arbitrarily uniform magnetization, which is a combination of the axial and diametrical magnetizations [23], can be predicted, thanks to the superposition principle.

These motivations led to the modelling and study of the magnetic field distribution of an elliptical cylinder with uniformly axial magnetization in this paper. The charge model [3] and geometric analyses are applied to derive the semi-analytical expressions of the axial, azimuthal, and radial components of the magnetic field created by the cylinder in three-dimensional (3D) space. FEA was conducted to verify the accuracy of the analytical model, and the efficiency of these expressions was also demonstrated. Furthermore, the fast-computed analytical model makes it possible to study the distribution of the magnetic field created by an axially magnetized cylinder with different elliptical profiles and compare it with that of a circular cylinder.

This paper is organized as follows. Section 2 describes the mathematical formulation of the analytical model. Section 3 presents the Finite Element verification. Section 4 discusses the distribution of the magnetic field of a cylinder with different elliptical profiles. Section 5 draws the conclusion.

2. MATHEMATICAL FORMULATION

The geometry of an elliptical cylinder used to derive expressions of the three magnetic field components is illustrated in Figure 1 and Figure 2. It is assumed that the magnetization \vec{J} is uniformly distributed along axis Z, as depicted in Figure 1(a), and the thickness of the cylinder is h. The elliptical profile of the cylinder is presented in Figure 2; here, F_1 and F_2 are the foci of the ellipse; a is the length of the semi-major axis; and b is the length of the semi-minor axis. The general equation for the ellipse Eq. (1) has a parametric representation using sine and cosine functions as $\vec{V} = (x, y) = (a \cos \tau, b \sin \tau), 0$ $\leq \tau < 2\pi; \tau$ can be defined as shown in Figure 2, based on de la Hire [24], where A is on a circle with a radius of a, and B is on a circle with a radius of b.

$$\frac{x^2}{a^2} + \frac{y^2}{b^2} = 1 \tag{1}$$



Figure 1. Computed elliptical cylinder: (a) 3D model of the elliptical cylinder; (b) Projection of the cylinder on an O'X'Y' plane.



Figure 2. Schematic of the elliptical profile of the cylinder [24].

Under static conditions, the well-known governing equations which are so-called Maxwell equations (Eqs. (M.1)-(M.4)) for electromagnetism can be presented as follows:

$$\vec{\nabla} \times \vec{H} = \vec{j} \tag{M.1}$$

$$\overline{\nabla} \cdot \overline{B} = 0 \tag{M.2}$$

$$\overrightarrow{\nabla} \times \overrightarrow{E} = 0$$
 (M.3)

$$\vec{\nabla} \cdot \vec{D} = \rho \tag{M.4}$$

and the constitutive equations linking \overrightarrow{B} to \overrightarrow{H} and \overrightarrow{D} to \overrightarrow{E} are expressed as $\overrightarrow{B} = \mu_0 \overrightarrow{H} + \overrightarrow{J}$ (M.5) and $\overrightarrow{D} = \varepsilon_0 \overrightarrow{E}$ (M.6), respectively.

Here, j (A/m²) and ρ (C/m³) are the electric current density and electric charge density, respectively; E is the electric field intensity (V/m); D is the electric flux density (C/m²); H is the magnetic field intensity (A/m); B is the magnetic flux density (T); μ_0 (H/m) and ε_0 (F/m) are the permeability and permittivity of the vacuum, respectively; J(T) is the intensity of magnetization.

By introducing the surface charge density σ_s and the volume charge density σ_v with some simplifications, these equations (Eqs. (M.1)–(M.6)) can be turned into a new form which is called the charge model (the derivation steps can be found in the work of Furlani [3].) to calculate the magnetic field created by a permanent magnet in three-dimensional free space. Based on this charge model, the magnetic field intensity $\overrightarrow{H_K}$ at any point $K(r, \alpha, z)$ in a cylindrical coordinate system can be computed using Eq. (2).

$$\overrightarrow{H_K} = \frac{1}{4\pi\mu_0} \left(\iint_s \frac{\sigma_s}{\left|\overrightarrow{PK}\right|^3} \overrightarrow{PK} ds + \iiint_v \frac{\sigma_v}{\left|\overrightarrow{PK}\right|^3} \overrightarrow{PK} dv \right)$$
(2)

where P is a point on the surface of the cylinder (Figure 1(a)); v and s denote the volume and surface area of the given magnet, respectively.

The volume charge density σ_v in Eq. (2) can be calculated as follows:

$$\sigma_v = - \vec{\nabla} \cdot \vec{J} \tag{3}$$

Since the magnetisation vector \overrightarrow{J} is uniformly axial, and its divergence is equal to zero, which means $\sigma_v = 0$. This means that the volume charge has no contribution to the magnetic field; hence, the magnetic field intensity $\overrightarrow{H_K}$ at point K can be calculated using only the surface charge:

$$\overrightarrow{H}_{K} = \frac{1}{4\pi\mu_{0}} \iint_{s} \frac{\sigma_{s}}{\left|\overrightarrow{PK}\right|^{3}} \overrightarrow{PK} ds \tag{4}$$

where the surface charge density σ_s can be calculated as follows:

$$\sigma_s = \overrightarrow{J} \cdot \overrightarrow{n} = \begin{cases} J & \text{for the upper surface} \\ 0 & \text{for the cylindrical surface} \\ -J & \text{for the lower surface} \end{cases}$$
(5)

Here, \overrightarrow{n} is the unit vector normal to the surface, for the top and bottom bases of the cylinder, and \overrightarrow{n} is parallel to vector \overrightarrow{J} and perpendicular to \overrightarrow{J} for the cylindrical surface. Thus, only the upper and lower surfaces of the cylinder contribute to the magnetic field.

From Eqs. (4) and (5), the magnetic field in a cylindrical coordinate system (r, α, z) with azimuth coincident with axis X is:

$$\vec{H}_{K(r,\alpha,z)} = \vec{H}_{K(r,\alpha,z)}^{+} + \vec{H}_{K(r,\alpha,z)}^{-}$$

$$= \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \int_{r_1=0}^{r_1=r_0} \frac{\overrightarrow{PK^+}}{\left|\overrightarrow{PK^+}\right|^3} r_1 dr_1 d\theta + \frac{-J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \int_{r_1=0}^{r_1=r_0} \frac{\overrightarrow{PK^-}}{\left|\overrightarrow{PK^-}\right|^3} r_1 dr_1 d\theta \qquad (6)$$

where the sign "+" refers to the magnetic field from the upper surface, and "-" refers to the lower surface.

From Figure 2,

$$r_0 = O'P_0 = \sqrt{a^2 \cos^2 \tau + b^2 \sin^2 \tau}$$
(7)

and

$$a\cos\tau = O'P_0\cos\theta \tag{8}$$

$$b\sin\tau = O'P_0\sin\theta \tag{9}$$

Dividing Eq. (8) by (9), we achieve:

$$\frac{a}{b}\cot\tau = \cot\theta \tag{10}$$

Inserting Eq. (10) into Eq. (7) produces:

$$r_0 = \frac{ab}{\sqrt{a^2 \sin^2 \theta + b^2 \cos^2 \theta}} \tag{11}$$

Firstly, we compute the magnetic field created by the upper surface.

From Figures 1(a) and 1(b), the vector $\overrightarrow{PK^+}$ can be projected on the radial, azimuthal, and axial directions $(\overrightarrow{u_r}, \overrightarrow{u_{\alpha}}, \text{ and } \overrightarrow{u_z})$ are the unit vectors respectively) as follows:

$$\overrightarrow{PK^{+}} = \overrightarrow{PP^{\prime}} + \overrightarrow{P'K^{\prime}} + \overrightarrow{K'K^{+}} = r_{1}\sin\left(\alpha - \theta\right)\overrightarrow{u_{\alpha}} + \left(r - r_{1}\cos\left(\alpha - \theta\right)\right)\overrightarrow{u_{r}} + \left(z - h\right)\overrightarrow{u_{z}}$$
(12)

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Inserting Eq. (12) into Eq. (6) gives:

$$\vec{H}_{K(r,\alpha,z)}^{+} = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \int_{r_1=0}^{r_1=r_0} \frac{r_1 \sin\left(\alpha-\theta\right) \vec{u}_{\alpha}^{+} + (r-r_1 \cos\left(\alpha-\theta\right)) \vec{u}_{r}^{+} + (z-h) \vec{u}_{z}^{+}}{\left(r_1^2 + r^2 - 2r_1 r \cos\left(\alpha-\theta\right) + (z-h)^2\right)^{\frac{3}{2}}} r_1 dr_1 d\theta \qquad (13)$$

The axial, azimuthal, and radial components of the magnetic field can be derived from Eq. (13), as detailed in the following sections.

2.1. The Axial Component $H^{\rm (3D)}_{K(z)}(r,\alpha,z)$

The axial component is expressed as the superposition of the contribution of two faces: $H_{K(z)}^{(3D)}(r, \alpha, z) = H_{K(z)}^{(3D)+}(r, \alpha, z) + H_{K(z)}^{(3D)-}(r, \alpha, z)$. From Eq. (13), $H_{K(z)}^{(3D)+}(r, \alpha, z)$ can be expressed as follows:

$$H_{K(z)}^{(3D)+}(r,\alpha,z) = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \int_{r_1=0}^{r_1=r_0} \frac{(z-h)}{\left(r_1^2 + r^2 - 2r_1r\cos\left(\alpha - \theta\right) + (z-h)^2\right)^{\frac{3}{2}}} r_1 dr_1 d\theta \qquad (14)$$

Integrating Eq. (14) based on r_1 with some simplifications yields (Appendix A):

$$H_{K(z)}^{(3D)+}(r,\alpha,z) = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \left(\frac{2\left(\delta r_0 - 2\xi_+\right)}{\left(4\xi_+ - \delta^2\right)\sqrt{r_0\left(r_0 - \delta\right) + \xi_+}} + \frac{4\sqrt{\xi_+}}{4\xi_+ - \delta^2} \right) (z-h) \, d\theta \tag{15}$$

where $\delta = 2r \cos(\alpha - \theta)$ and $\xi_{+} = r^{2} + (z - h)^{2}_{2}$.

It is noted that $4\xi_+ - \delta^2 = 4r^2 + 4(z-h)^2 - 4r^2\cos^2(\alpha-\theta) > 0$ except the singularity case when z = h and $\alpha - \theta = l\pi$; $l = 0, 1, 2, 3, \ldots$ This singularity can be easily removed when executing Eq. (15) to avoid indefinite result.

The contribution of the lower surface can be derived, following the same steps as above to achieve, with $\xi_{-} = r^2 + z^2$. Similar to the above notation, $4\xi_{-} - \delta^2 = 4r^2 + 4z^2 - 4r^2\cos^2(\alpha - \theta) > 0$ except the singularity case when z = 0 and $\alpha - \theta = l\pi$; l = 0, 1, 2, 3, ...

$$H_{K(z)}^{(3D)-}(r,\alpha,z) = \frac{-J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \left(\frac{2\left(\delta r_0 - 2\xi_{-}\right)}{\left(4\xi_{-} - \delta^2\right)\sqrt{r_0\left(r_0 - \delta\right) + \xi_{-}}} + \frac{4\sqrt{\xi_{-}}}{4\xi_{-} - \delta^2} \right) (z) \, d\theta \tag{16}$$

2.2. The Tangential (Azimuthal) Component $H_{K(\alpha)}^{(3D)}(r, \alpha, z)$

The azimuthal component $H_{K(\alpha)}^{(3D)}(r,\alpha,z) = H_{K(\alpha)}^{(3D)+}(r,\alpha,z) + H_{K(\alpha)}^{(3D)-}(r,\alpha,z)$. From Eq. (13), $H_{K(\alpha)}^{(3D)+}(r,\alpha,z)$ can be expressed as follows:

$$H_{K(\alpha)}^{(3D)+}(r,\alpha,z) = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \int_{r_1=0}^{r_1=r_0} \frac{r_1^2 \sin\left(\alpha-\theta\right)}{\left(r_1^2 + r^2 - 2r_1 r \cos\left(\alpha-\theta\right) + (z-h)^2\right)^{\frac{3}{2}}} dr_1 d\theta \tag{17}$$

Integrating Eq. (17) based on r_1 with some simplifications produces (Appendix A):

$$H_{K(\alpha)}^{(3D)+}(r,\alpha,z) = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \left(\ln\left(2\left(\sqrt{r_0(r_0-\delta)+\xi_+}+r_0\right)-\delta\right) - \frac{2\left(2\xi_+-\delta^2\right)r_0+2\delta\xi_+}{(4\xi_+-\delta^2)\sqrt{r_0(r_0-\delta)+\xi_+}} - \ln(2\sqrt{\xi_+}-\delta) + \frac{2\delta\xi_+}{(4\xi_+-\delta^2)\sqrt{\xi_+}}\right) \sin(\alpha-\theta)d\theta$$
(18)

The contribution of the lower surface can be derived, following the same steps as above to achieve:

$$H_{K(\alpha)}^{(3D)-}(r,\alpha,z) = \frac{-J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \left(\ln\left(2\left(\sqrt{r_0(r_0-\delta)+\xi_-}+r_0\right)-\delta\right) - \frac{2\left(2\xi_--\delta^2\right)r_0+2\delta\xi_-}{(4\xi_--\delta^2)\sqrt{r_0(r_0-\delta)+\xi_-}} - \ln(2(\sqrt{\xi_-})-\delta) + \frac{2\delta\xi_-}{(4\xi_--\delta^2)\sqrt{\xi_-}}\right) \sin(\alpha-\theta)d\theta$$
(19)

2.3. The Radial Component $H_{K(r)}^{(3D)}(r, \alpha, z)$

The radial component $H_{K(r)}^{(3D)}(r, \alpha, z) = H_{K(r)}^{(3D)+}(r, \alpha, z) + H_{K(r)}^{(3D)-}(r, \alpha, z)$. From Eq. (13), $H_{K(r)}^{(3D)}(r, \alpha, z)$ can be expressed as follows:

$$H_{K(r)}^{(3D)}(r,\alpha,z) = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \int_{r_1=0}^{r_1=r_0} \frac{(r-r_1\cos\left(\alpha-\theta\right))}{\left(r_1^2+r^2-2r_1r\cos\left(\alpha-\theta\right)+(z-h)^2\right)^{\frac{3}{2}}} r_1 dr_1 d\theta \tag{20}$$

Integrating Eq. (20) based on r_1 with $\Upsilon = \cos(\alpha - \theta)$ with some simplifications gives (Appendix A):

$$H_{K(r)}^{(3D)+}(r,\alpha,z) = \frac{J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \left(\frac{2\left(\delta r + 2\xi_+ \Upsilon - \delta^2 \Upsilon\right) r_0 - 4\xi_+ r + 2\delta\xi_+ \Upsilon}{(4\xi_+ - \delta^2)\sqrt{r_0(r_0 - \delta) + \xi_+}} - \Upsilon \ln\left(2\left(\sqrt{r_0(r_0 - \delta) + \xi_+} + r_0\right) - \delta\right) - \frac{-4\xi_+ r + 2\delta\xi_+ \Upsilon}{(4\xi_+ - \delta^2)\sqrt{\xi_+}} + \Upsilon \ln(2\sqrt{\xi_+} - \delta))d\theta(21)\right)$$

The contribution of the lower surface can be derived, following the same steps as above to achieve:

$$H_{K(r)}^{(3D)-}(r,\alpha,z) = \frac{-J}{4\pi\mu_0} \int_{\theta=-\pi}^{\theta=\pi} \left(\frac{2\left(\delta r + 2\xi_-\Upsilon - \delta^2\Upsilon\right)r_0 - 4\xi_-r + 2\delta\xi_-\Upsilon}{(4\xi_- - \delta^2)\sqrt{r_0(r_0 - \delta) + \xi_-}} -\Upsilon\ln\left(2\left(\sqrt{r_0(r_0 - \delta) + \xi_-} + r_0\right) - \delta\right) - \frac{-4\xi_-r + 2\delta\xi_-\Upsilon}{(4\xi_- - \delta^2)\sqrt{\xi_-}} + \Upsilon\ln(2\sqrt{\xi_-} - \delta))d\theta (22)$$

From the magnetic field intensity \overrightarrow{H}_K , the magnetic flux density \overrightarrow{B}_K can be computed with:

$$\vec{B}_K = \mu_0 \vec{H}_K \qquad \text{(in the air space)} \qquad (23)$$

$$\vec{B}_K = \mu_0 \vec{H}_K + \vec{J}$$
 (inside the magnet) (24)

In the case where a = b, the elliptical cylinder is simplified to a circular cylinder; hence the derived expressions could also be implemented to calculate the magnetic field generated by a circular cylinder permanent magnet with uniformly axial magnetization.

Following the same derivation steps as presented above, the magnetic field created by an arc-shaped elliptical cylinder (which is aligned with the Cartesian and cylindrical coordinate systems as presented in Figure 3) with uniformly axial magnetization can be computed using the derived expressions for the three components with an integral interval of θ from 0 to the arc angle γ instead of $[-\pi, \pi]$. The calculation of magnetic field of this arc-shaped cylinder can facilitate the design and optimization of a Hallbach permanent magnet [25] which may include pieces of permanent magnet of this kind.

3. FINITE ELEMENT VERIFICATION

Finite Element Analysis (FEA) is a commonly used method to study problems in various fields. In static electromagnetism, this method aims at solving the governing and constitutive equations listed in Section 2 (Eqs. (M.1)–(M.6)). In this research, FEA is implemented to verify the accuracy of the semi-analytical model using Electromagnetic Simulation Software®(EMS) (EMWorks, Inc., Montreal, Quebec, Canada). The parameters of the used elliptical cylinder made of rare earth material (Figure 2) are: a = 6 mm, b = 3 mm, h = 5 mm, and magnetization J = 1 T which is axially oriented.



Figure 3. Arc-shaped elliptical cylinder.



Figure 4. FEA verification results for axial component (the parameters of the used elliptical cylinder (Figure 2) are: a = 0.006 m, b = 0.003 m, h = 0.005 m, J = 1 T); plots along the radial distance with z = 0.007 m and r from 0 m to 0.020 m.



Figure 5. FEA verification results for radial component (the parameters of the used elliptical cylinder (Figure 2) are: a = 0.006 m, b = 0.003 m, h = 0.005 m, J = 1 T); plots along the radial distance with z = 0.007 m and r from 0 m to 0.020 m.



Figure 6. FEA verification results for azimuthal component (the parameters of the used elliptical cylinder (Figure 2) are: a = 0.006 m, b = 0.003 m, h = 0.005 m, J = 1 T); plots along the radial distance with z = 0.007 m and r from 0 m to 0.020 m.

The plots of three components of the magnetic field generated by the given elliptical cylinder are illustrated in Figures 4, 5, and 6. Three components of magnetic field of ten points with random r (mm) are extracted from these figures and listed in Table 1 including the errors of computed results using the derived analytical model against those of Finite Element Analysis.

Table 1.	Analytical	results vs	FEA re	esults for	10 points	with r	andom r	(mm)	and	errors	of r	esults
between tw	vo methods	(Max -	Maximu	ım; Aver -	- Average	e; Min	— Minim	im).				

Points																
(r mm,	Axial component			Azimuthal component				Radial component								
30°, 7	(Gauss)			(Gauss)			(Gauss)									
mm)																
	Analytical		FEA		Analytical		FEA results		Analytical		FEA results					
	resul	ts	r	esults	results		results		T L/Y TCSUItS							
1.01	2004.985		20	012.765	142.500		147.123		282.559		286.766					
2.22	1828.209		18	329.148	302.959		30	9.872	623.767		633.473					
3.43	1457.334		14	57.554	417.049		411.765		931.164		933.559					
5.25	599.604		596.804		350.571		348.170		1019.055		1018.313					
6.67	145.592		144.049		185.266		184.216		733.565		730.477					
8.89	-58.102		-58.166		54.862		54.466		348.896		348.724					
13.13	-59.326		-	59.374	7.207		7.217		92.435		92.307					
17.17	-34.112		-34.081		1.579		1.563		33.517		33.484					
18.18	-29.717		-29.686		1.135		1.133		26.859		26.834					
19.79	-24.015		-23.986		0.693		0.687		19.289		19.223					
Error	Max	Ave	er	Min	Max	А	ver Min		ver Min		ver Min		Max	Av	ver	Min
(%)	1.07	0.2	5	0.015	3.14	1	.08	0.13	1.53	0.4	45	0.049				

It can be seen from Table 1 and Figures 4, 5, and 6 that the results of the magnetic field computed using the derived semi-analytical model are in excellent agreement with those computed using the FEA model (the average errors between two models are 0.25% for axial component, 1.08% for azimuthal component, and 0.45% for radial component (Table 1)).

Evaluated in MATLAB R2016b (MathWorks, Natick, MA, USA), it took an average of 0.75 ms on a personal computer (Intel Core i7-6700 CPU, 3.40 GHz) to calculate each component, using the derived expressions at a single location (2000 samplepoints were used in this evaluation), whereas on the other hand, it could take up to several hours with small scale of mesh (the smaller the mesh is, the more computation time it consumes) to complete the simulation using FEA on the same personal computer.

4. ANALYSIS OF THE MAGNETIC FIELD DISTRIBUTION

In this section, the magnetic field distribution created by an axially magnetized cylinder with different elliptical profiles is plotted along the radial distance r in the interval [0 mm, 20 mm] with the azimuthal angle $\alpha = \pi/6$, and along the azimuthal angle α (alpha) in the interval of $[-\pi, \pi]$ with the radial distance r = 8 mm. The thickness of the used cylinder is h = 5 mm; the magnetic remanence J = 1 T; the axial coordinate of the computed profile is z = 7 mm; the elliptical profiles of the cylinder vary due to the



Figure 7. Distribution of the axial component (the thickness of the used cylinder is h = 5 mm; the magnetic remanence J = 1 T; the axial coordinate of the computed profile is z = 7 mm): (a) plot along the radial distancer in the interval [0 mm, 20 mm] with azimuthal angle $\alpha = \pi/6$; (b) plot along the azimuthal angle α (alpha) with radial distance r = 8 mm.

changes in the so-called axis ratio c of the major semi-axis a and semi-minor axis b, c = a/b. The plots of the magnetic field distribution were carried out in six cases: c1 = 6/3 = 2/1; $c2 = \sqrt{18}/\sqrt{18} = 1/1$; c3 = 9/2; c4 = 18/1; c5 = 2/9 and c6 = 1/18. It is noted here that regardless of the variations in profiles, the ellipses have the same area $S = \pi ab = 18\pi \text{ mm}^2$.

Figure 7(a) shows that the amplitude of the axial component of a circular cylinder (c2 = 1/1) is the largest (2143 Gauss) at the beginning r = 0 mm, and it eventually drops down to below those of c1 (from r = 2.8 mm), c3 (from r = 4.5 mm) and c4 (from r = 6.1 mm). All the other amplitudes decrease when r increases. Figure 7(b) demonstrates that the magnetic field of a circular cylinder (c2 = 1/1) is a negative constant (-57.64 Gauss) along the azimuthal angle (Table 2), whereas the magnetic field distribution of an elliptical cylinder is periodic and obtains both positive and negative values; the amplitudes of the periodical distribution (Table 2) are 62.69 Gauss for c1, 810.0 Gauss for c3, 963.1 Gauss for c4, 810.0 Gauss for c5, and 963.2 Gauss for c6. It can be concluded that, for the same area, the elliptical cylinder, compared with the circular cylinder, has the interesting properties of producing larger magnetic field amplitudes and periodical distributions of the magnetic field.

Figures 8(a) and 8(b) show that the azimuthal distribution of the magnetic field of a circular cylinder is zero (Table 2) due to its symmetry. However, elliptical cylinders can yield large amplitudes, which vary according to changes of the axis ratio c. The magnetic field distribution along the radial distance r can obtain both negative and positive values; their maximum amplitudes are 432.9 Gauss at r = 4 mm for c1, 683.8 Gauss at r = 4 mm for c3, 522.2 Gauss at r = 3.8 mm for c4, -433.8 Gauss at r = 2.7 mm for c5 and -306.1 Gauss at r = 2.3 mm for c6. The magnetic field distribution of the



Figure 8. Distribution of the azimuthal component (the thickness of the used cylinder is h = 5 mm; the magnetic remanence J = 1 T; the axial coordinate of the computed profile is z = 7 mm): (a) plot along the radial distance r in the interval [0 mm, 20 mm] with azimuthal angle $\alpha = \pi/6$; (b) plot along the azimuthal angle α (alpha) with radial distance r = 8 mm.

Arria notio $a - a/b$	Axial component	Azimuthal component	Radial component			
Axis fatio $c = a/b$	(Gauss)	(Gauss)	(Gauss)			
c1 = 2/1	62.69	89.45	594.0			
c2 = 1/1	-57.64	0	411.7			
c3 = 9/2	810.0	384.2	756.2			
c4 = 18/1	963.1	536.0	248.6			
c5 = 2/9	810.0	384.2	756.2			
c6 = 1/18	963.2	536.0	248.6			

Table 2. Amplitudes of the periodical distribution of magnetic field.

azimuthal component (Table 2) is periodical along the azimuthal angle (Figure 8(b)); the amplitudes of this distribution are 89.45 Gauss for c1, 384.2 Gauss for c3, 536.0 Gauss for c4, 384.2 Gauss for c5, and 536.0 Gauss for c6.

In contrast with the distribution of axial and azimuthal components, the radial component of the field consists of only positive values (Figures 9(a) and 9(b)). The amplitudes of the radial component of a circular and an elliptical cylinder vary according to the changes of r; the maximum amplitudes are 1065 Gauss at r = 4.6 mm for c1, 1224 Gauss at r = 4.1 mm for c2, 725.4 Gauss at r = 4.8 mm for c3,



Figure 9. Distribution of the radial component (the thickness of the used cylinder is h = 5 mm; the magnetic remanence J = 1 T; the axial coordinate of the computed profile is z = 7 mm): (a) plot along the radial distance r in the interval [0 mm, 20 mm] with azimuthal angle $\alpha = \pi/6$; (b) plot along the azimuthal angle α (alpha) with radial distance r = 8 mm.

341.4 Gauss at r = 3.9 mm for c4, 947 Gauss at r = 2.8 mm for c5, and 553.3 Gauss at r = 2.3 mm for c6. The value of this component of the circular cylinder (c2 = 1/1) is a positive constant (411.7 Gauss) along the azimuthal angle (Table 2). The distribution of this component of an elliptical cylinder along the azimuthal angle is periodical; however, its amplitudes and phases change due to the variations of c. For the same area, elliptical cylinders with c1 = 2/1, c3 = 9/2, and c5 = 2/9 can produce the periodical magnetic field with larger amplitudes than a circular cylinder. The amplitudes of the periodical distribution (Table 2) are 594.0 Gauss for c1, 756.2 Gauss for c3, 248.6 Gauss for c4 and c6, and 756.2 Gauss for c5. The variations in magnetic field strength of axial, azimuthal and radial components may be used in novel sensing applications.

5. CONCLUSION

In this research, the semi-analytical expressions of three components of the magnetic field created by an elliptical cylinder with axial magnetization in three dimensional space are derived. The magnetic field distribution created by the permanent magnet is analysed. The results of the new expressions are in excellent agreement with those of FEA. These expressions are fast-computed and can be extended to compute the magnetic field created by an arc-shaped elliptical cylinder with axial magnetization which can be used in a novel Halbach array as well as a circular cylinder with the same magnetization direction. The magnetic field distribution of an elliptical cylinder with axial magnetization along the azimuthal angle is periodical whereas a circular cylinder produces a constant value. Along the radial distance, the axial and radial components of the magnetic field of a circular cylinder can obtain larger values, but eventually drop down over the counterparts of an elliptical cylinder. The azimuthal component of a circular cylinder is zero due to its symmetry whereas an elliptical cylinder produces large amplitudes.

APPENDIX A. INTEGRATION STEP

This section presents the step integration used when integrating Eqs. (14), (17), and (20). For the purpose of generality, the following step integration is carried out for two key components (Eqs. (A1) and (A2)) in the mentioned equations. The integral variables in these equations are replaced by "t" for briefness.

A.1. Step Integration of $A = \int \frac{tdt}{(t^2 - at + b)^{\frac{3}{2}}}$ with $4b > a^2$

$$A = \int \frac{tdt}{(t^2 - at + b)^{\frac{3}{2}}} = \int \frac{tdt}{\left(\left(t - \frac{a}{2}\right)^2 + \frac{4b - a^2}{2}\right)^{\frac{3}{2}}} = \int \frac{\left(t + \frac{a}{2}\right)dt}{\left(\left(t - \frac{a}{2}\right)^2 + \frac{4b - a^2}{2}\right)^{\frac{3}{2}}} - \int \frac{\frac{a}{2}dt}{\left(\left(t - \frac{a}{2}\right)^2 + \frac{4b - a^2}{2}\right)^{\frac{3}{2}}}$$
(A1)

Denote $u = t - \frac{a}{2}$, Eq. (A1) becomes

$$A = \int \frac{u du}{\left((u)^2 + \frac{4b - a^2}{2}\right)^{\frac{3}{2}}} - \int \frac{\frac{a}{2} du}{\left((u)^2 + \frac{4b - a^2}{2}\right)^{\frac{3}{2}}}$$
(A2)

Integrating the basic integrals of Eq. (A2) based on u and then returning the variable to t with some simplifications gives:

$$A = \frac{2(at-2b)}{(4b-a^2)\sqrt{t(t-a)+b}} + C$$
(A3)

Here C is the integral constant.

A.2. Step Integration of $B = \int \frac{t^2 dt}{(t^2 - at + b)^{\frac{3}{2}}}$ with $4b > a^2$

$$B = \int \frac{(t^2 - at + b) dt}{(t^2 - at + b)^{\frac{3}{2}}} - \int \frac{atdt}{(t^2 - at + b)^{\frac{3}{2}}} + \int \frac{bdt}{(t^2 - at + b)^{\frac{3}{2}}}$$
(A4)

Denote $u = t - \frac{a}{2}$, Eq. (A4) becomes

$$B = \int \frac{du}{\sqrt{(u)^2 + \frac{4b-a^2}{2}}} - aA + \int \frac{bdu}{\left((u)^2 + \frac{4b-a^2}{2}\right)^{\frac{3}{2}}}$$
(A5)

Integrating the basic integrals of Eq. (A5) and applying the result of the above integration, then returning to variable t with some simplifications produces:

$$B = \ln\left(\left|2\left(\sqrt{t(t-a)+b}+t\right) - a\right|\right) - \frac{2t(2b-a^2)+2ab}{(4b-a^2)\sqrt{t(t-a)+b}} + C$$
(A6)

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