

Compact High Isolation Meandered-Line PIFA Antenna for LTE (Band-Class-13) Handset Applications

Ahmed A. Naser¹, Khalil H. Sayidmarie^{2, *}, and Jabir S. Aziz³

Abstract—MIMO systems have become an essential part in many communications networks and Long Term Evolution (4G) mobile communication systems. Mobile handsets using lower band of LTE (LTE-700 band) require antennas of reduced size that can be adapted to the limited space in the handset. This paper presents the design, optimization and implementation of two meandered-line PIFA antennas working as an MIMO system with high isolation for LTE-700 band mobile applications. To solve the problem of mutual coupling, a combination of decoupling arrangements was used to improve the isolation between the two antennas. The influences of various design parameters are investigated using the CST Microwave Studio Suite. A prototype of the proposed Meandered-line PIFA Antenna was fabricated and tested using vector network analyzer. Good agreement was found between the simulated and measured results. The fabricated MIMO antenna shows an isolation better than 12 dB and a -6 dB bandwidth of (75 MHz) in the frequency range from (720 MHz) to (795 MHz). The antenna has 1.94 dB gain, total efficiency of 85%, and volume of $110 \times 65 \times 1.6$ mm³, that is $(0.275 \times 0.1625 \times 0.004)$ in wavelengths.

1. INTRODUCTION

The Long-Term Evolution (LTE) is a broadband solution that offers various features with good flexibility in terms of deployment options and potentials for mobile services [1]. Recent wireless communication systems require higher data rates to support the increasing mobile services [2]. An MIMO system can increase the channel capacity and improves reliability without sacrificing spectrum efficiency or consuming additional transmitted power [3]. MIMO systems constructively explore multi-path propagation using different transmission paths to the receiver to improve wireless systems capacity, range and reliability [3]. Various designs of MIMO antennas have been reported for the mobile handset using PIFA and monopole configurations [4–8].

In an MIMO antenna system, high antenna coupling results from signal leakage from one antenna to another leading to reduced antenna efficiency due to the power absorbed by a close element with strong coupling. The problem is magnified in the portable wireless devices where the MIMO antennas are placed very close due to the limited space. Therefore, the reduction of coupling is a major consideration in MIMO antenna design to reduce the correlation between the multiple elements.

Significant research efforts have been reported to reduce the mutual coupling [6–13]. A T-shape slot impedance transformer added to both single and dual band PIFA's resulted in an isolation better than 20 dB [6]. In [7] an additional non-radiating folded shorting strip connected between the antenna element and ground plane resulted in an isolation better than 28 dB for the lower WLAN band. In [8], two quarter-wavelength slots in the ground plane were used to reduce the coupling. Slits were also introduced in the ground plane to reduce mutual coupling and enhance isolation of the MIMO antenna [9]. Insertion

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* Corresponding author: Khalil Hassan Sayidmarie (kh.sayidmarie@gmail.com).

¹ Electrical Engineering Department, College of Engineering, University of Mosul, Iraq. ² Department of Communication Engineering, College of Electronic Engineering, Ninevah University, Mosul, Iraq. ³ College of Engineering, University of Al-Nahrain, Baghdad, Iraq.

of a neutralization line between the two antenna elements is a promising method that has been used in [10, 11].

A decoupling network consisting of two sections of a transmission line is placed at the center of the two antenna elements in an LTE MIMO system is recently proposed [12]. The isolation was enhanced by approximately 23 dB comparing to that without a decoupling network. The fabricated MIMO antenna satisfied a 6-dB return loss requirement and the envelope correlation coefficient is maintained below 0.4 over the LTE band 13 ranging from 746 to 787 MHz [12]. A decoupling element for enhancing the port isolation in MIMO antennas for triple (LTE) bands was presented in [13]. A spiral decoupling structure is added between two strongly coupled antennas to enhance the port isolation by 20 dB [13]. In [14] a fork-shaped structure was introduced in the ground plane to decrease the mutual coupling between the antennas. The obtained simulation results show a high isolation of about 20 dB. The reduction of antenna coupling in MIMO systems becomes a challenging problem at the lower band of the LTE since the allowable separation is a very small fraction of the wavelength.

In this paper, an MIMO meandered-line PIFA antenna is proposed for LTE-700 MHz band applications. The meandered-line can collect longer electrical length in a given physical area and thus helps to reduce the size of the antenna. This feature is desired since the antenna operates at low frequencies of the LTE Band 13. By combining the meandered-line with the PIFA structure, their desired properties are integrated. A combination of decoupling arrangements are investigated to improve the isolation between the two antennas. The parameters of the isolation techniques were optimized by using Particle Swarm Optimization (PSO) and Nelder-Mead optimization techniques for further improvement of the performance. These techniques are built in the CST Microwave Studio Suite which was used to analyze and optimize the proposed antenna. The proposed MIMO antennas configuration is compact and can be fitted in conventional mobile handsets.

2. DESIGN AND SIMULATION OF MIMO MEANDERED-LINE PIFA ANTENNAS

The proposed design consists of two meandered-line PIFA antennas configuration using the same substrate as shown in Fig. 1. Each antenna is based on the recently published antenna design [15]. The antenna combines a meandered-line which is known in its compact features and PIFA configuration to achieve the desired features. The antenna shape was optimized to achieve a bandwidth of (58) MHz centered at (765) MHz and VSWR of (1.04). The antenna has a compact volume of $(65 \times 110 \times 1.6) \text{ mm}^3$ that is a $(0.165 \times 0.28 \times 0.004)$ wavelength at the center frequency of the lower band). This antenna was considered as a primary antenna for the MIMO antenna system proposed in this paper.

Figure 2 presents the simulation results of the MIMO configuration, which shows a lower band

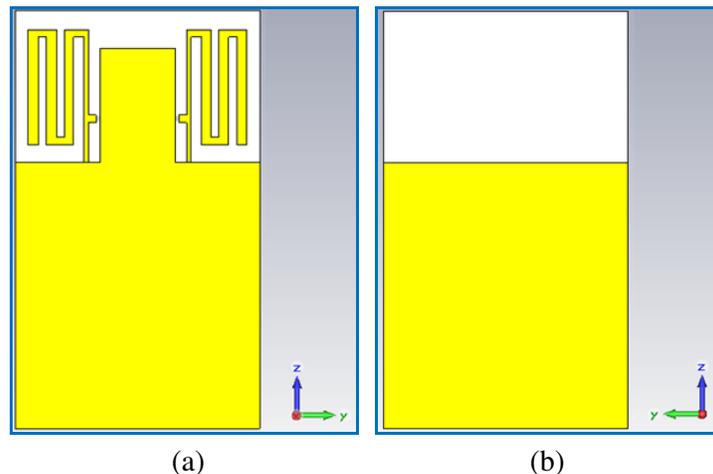


Figure 1. Two optimized meandered-line PIFA antennas, (a) (top layer) and (b) ground (bottom layer).

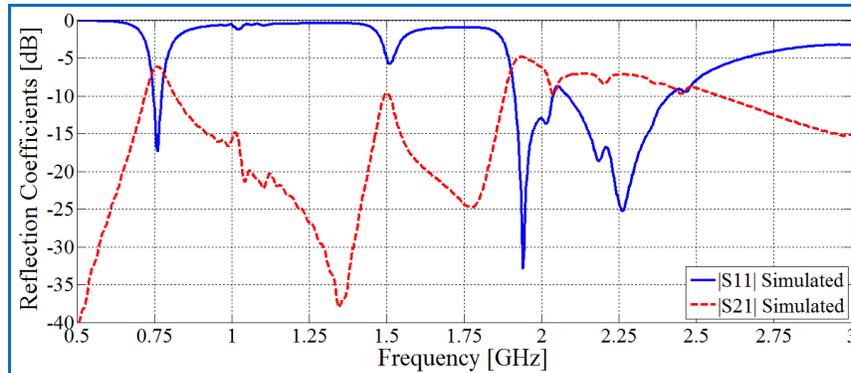


Figure 2. Variation of the reflection and coupling coefficients with frequency for the antennas in Fig. 1.

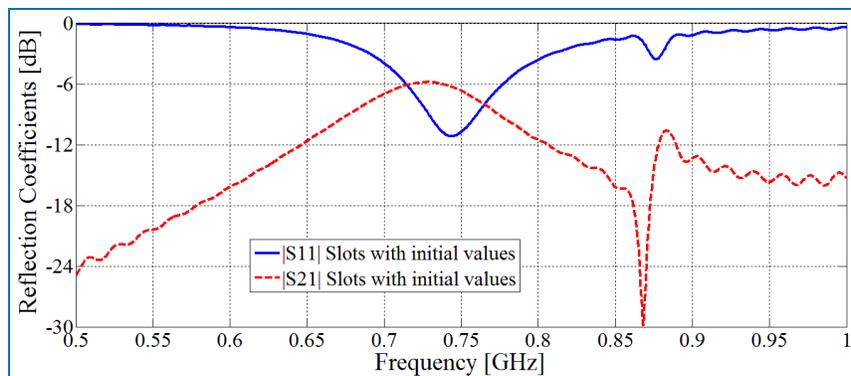


Figure 3. Reflection and coupling coefficients of the antennas when only slots were used.

around 760 MHz having a -6 dB bandwidth of (52) MHz. There is a second band across the range (1.9–2.52) GHz. The mutual coupling between the two antennas is relatively high at about (-6 dB). Hence, there is a need to implement some decoupling arrangements that can decouple the two antennas to improve the ports isolation and meet the design criteria.

3. ISOLATION ENHANCEMENT OF THE PROPOSED MIMO SYSTEM

The decoupling arrangement used in this paper is a combination of the following techniques in an attempt to integrate their desired features:

- Slot technique.
- Neutralization line technique.
- Fork-shaped structure in the ground plane.

Figures 3 through 8 show the effect of each of the used decoupling techniques on the performance of the MIMO antennas, where all the simulation results were obtained with the initial parameters values (see Table 1). It can be seen from Figs. 3 through 5 that using the neutralization line technique is the most effective compared to the slots and fork-shaped line techniques in the reduction of mutual coupling. Merging of the two techniques was also introduced (see Figs. 6 to 8). However, the results of combining any two configurations were not better than that of the neutralization line technique.

A proposed design using the three decoupling techniques was investigated in an attempt to collect the advantages of the three techniques. Fig. 9 shows the structure of the proposed meandered-line PIFA antennas with the three implemented decoupling techniques. The performance of the proposed antenna cannot be easily optimized manually. Therefore, an optimization process using Particle Swarm

Table 1. Initial values of the MIMO antenna parameters.

Parameter	Initial value [mm]	Parameter	Initial value [mm]
The antenna parameters		<i>ls1</i>	25
<i>a</i>	32.4	<i>ws</i>	1
<i>g1</i>	2	<i>ws1</i>	1
<i>g2</i>	2	The N-line parameters	
<i>g3</i>	3	<i>wn</i>	0.5
<i>gap</i>	26.5	<i>gnh</i>	0.5
<i>ha</i>	5.25	<i>gnv</i>	4.5
<i>w1</i>	1.1	The fork parameters	
<i>w2</i>	2	<i>lfo1</i>	30
<i>w3</i>	2	<i>lfo2</i>	7
<i>w4</i>	2	<i>wfo</i>	20
The slot parameters		<i>wfoh</i>	1
<i>ls</i>	30	<i>wfov</i>	2

Antenna parameters
 Slot parameters
 N-line parameters
 Fork parameters

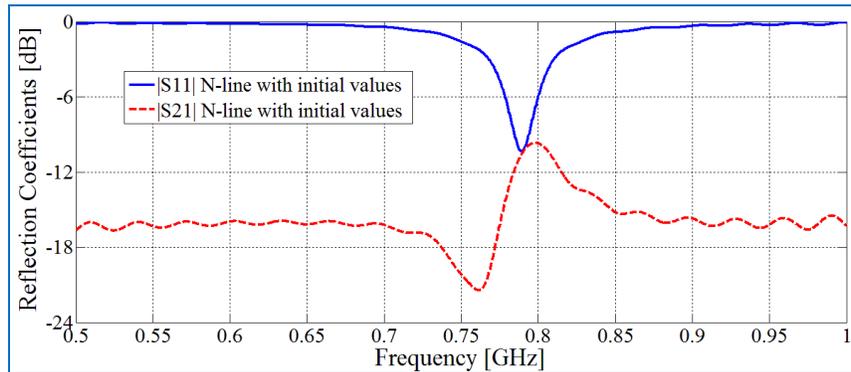


Figure 4. Reflection and coupling coefficients of the antennas when only N-line was used.

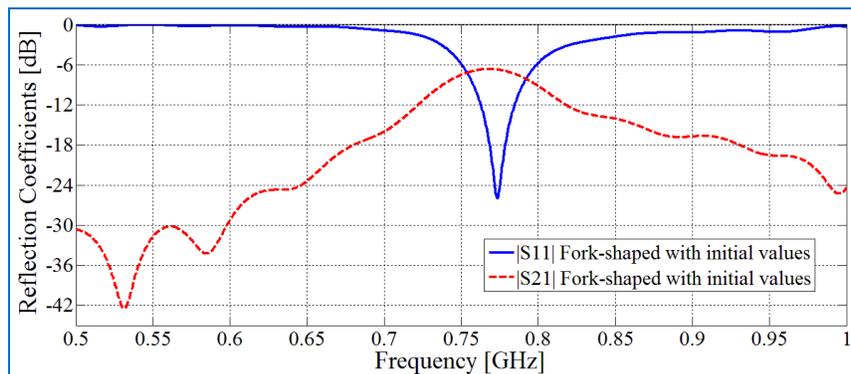


Figure 5. Reflection and coupling coefficients of the antennas when only the fork-shaped line was used.

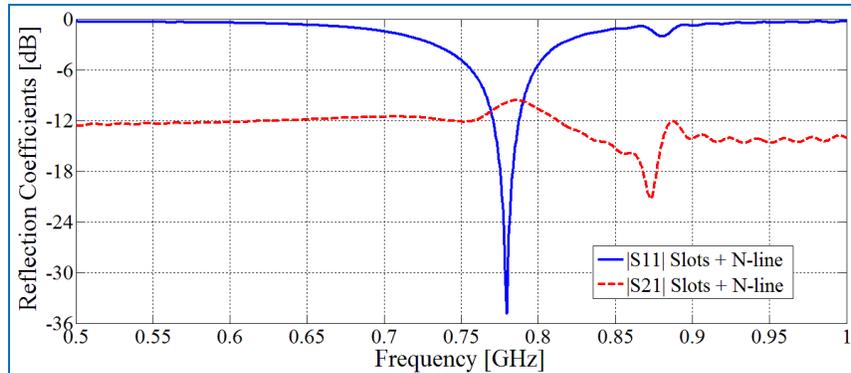


Figure 6. Reflection and coupling coefficients of the antennas when both slots and N-line were used.

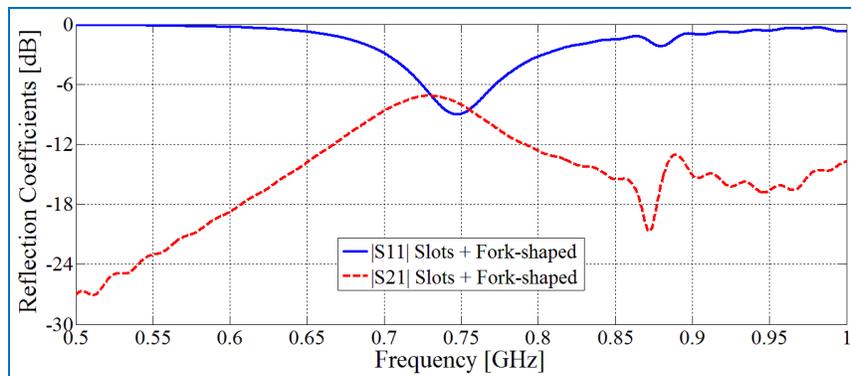


Figure 7. Reflection and coupling coefficients of the antennas when both slots and fork-shaped line were used.

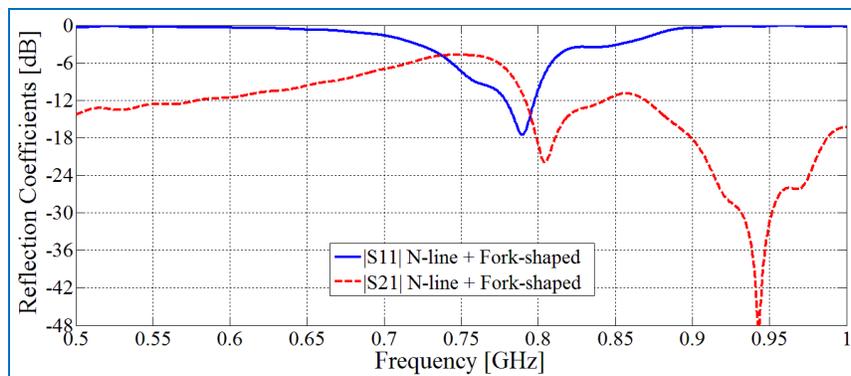


Figure 8. Reflection and coupling coefficients of the antennas when both N-line and fork-shaped line were used.

Optimization (PSO) and Nelder-Mead optimization were used to optimize the parameters of the antenna structure and the decoupling arrangements. These two techniques are already implemented in the CST Microwave Studio Suite. Fig. 10 shows a detailed list of the parameters of each part of the antenna designs; slots, neutralization-line and fork-shaped structure. The optimization process also took into account the values of the basic antenna parameters. Table 2 shows the initial and optimized parameter values of the proposed MIMO antenna.

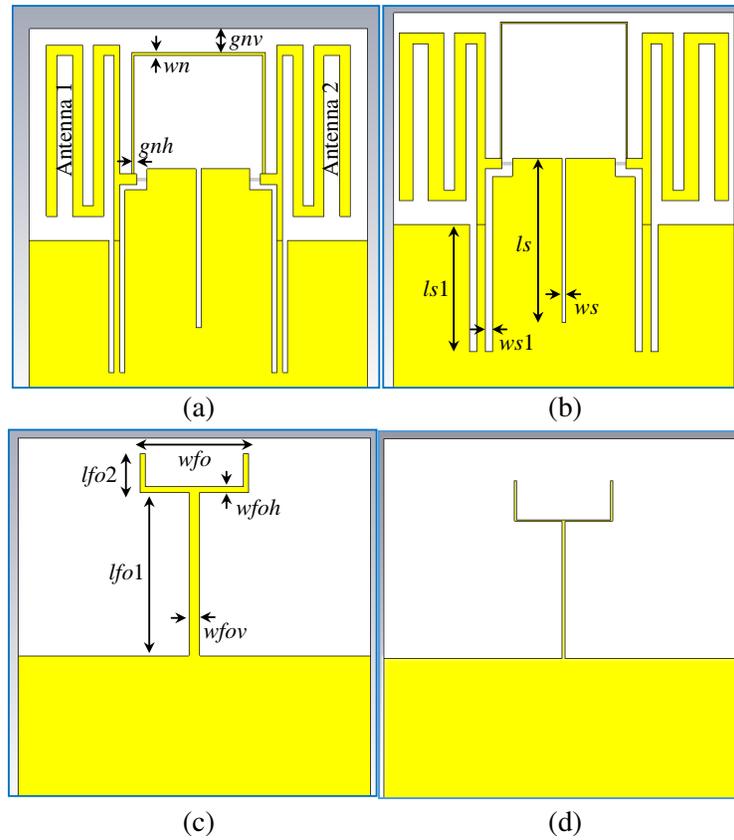


Figure 9. Two meandered-line PIFA antennas, (a) and (b) are the initial and the optimized antenna (top layer) respectively, (c) and (d) represent the initial and the optimized ground (bottom layer) respectively. Dimensions are given in Table 2.

Table 2. Initial and optimized values of the parameters of the MIMO Meandered-Line PIFA of Fig. 10.

Parameter	Value (mm)		Parameter	Value (mm)	
	initial	optimized		initial	optimized
The antenna parameters			$ls1$	25	24
a	32.4	31.7	ws	1	0.6
$g1$	2	2	$ws1$	1	1.4
$g2$	2	2	The neutralization-line parameters		
$g3$	3	3	wn	0.5	0.3
gap	26.5	27.5	gnh	0.5	0
ha	5.25	5.2	gnv	4.5	1.7
$w1$	1.1	1.6	The fork-shaped parameters		
$w2$	2	2.25	$lfo1$	30	25
$w3$	2	3	$lfo2$	7	7.4
$w4$	2	2.5	wfo	20	18.5
The slot parameters			$wfoh$	1	0.3
ls	30	31	$wfov$	2	0.6

Antenna parameters
 Slot parameters
 N-line parameters
 Fork parameters

4. SIMULATION RESULTS

Figure 11 shows the reflection and coupling coefficients of the proposed antenna with the three decoupling techniques shown in Figs. 9 and 10. The -6 dB bandwidth is (52 MHz) covering the range (741–793) MHz. The coupling across this band is between $(-11$ and $-12)$ dB showing an improvement of 6 dB.

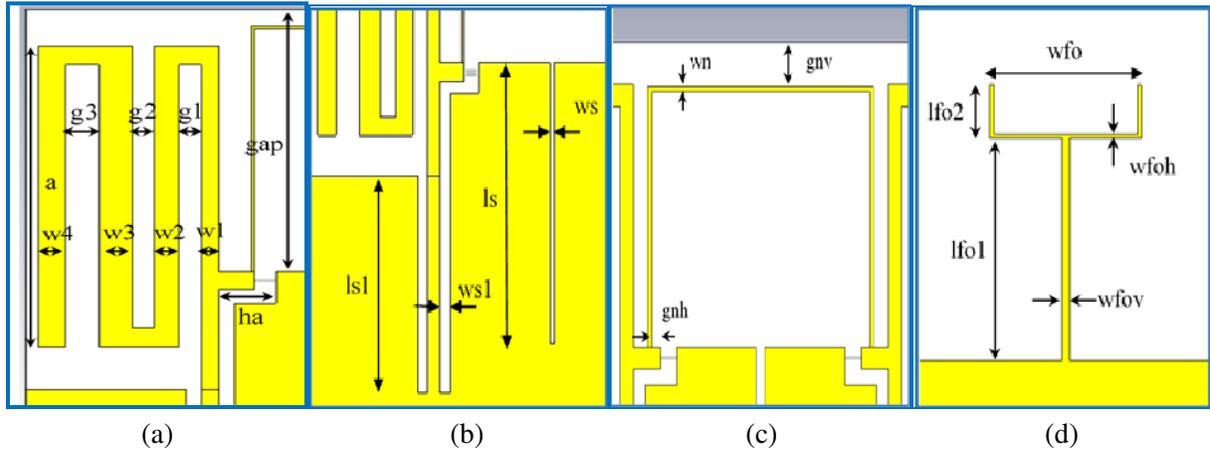


Figure 10. Parameters of the final PIFA structure; (a) antenna parameters, (b) slot parameters, (c) neutralization-line parameters and (d) fork-shaped line parameters. Dimensions are given in Table 2.

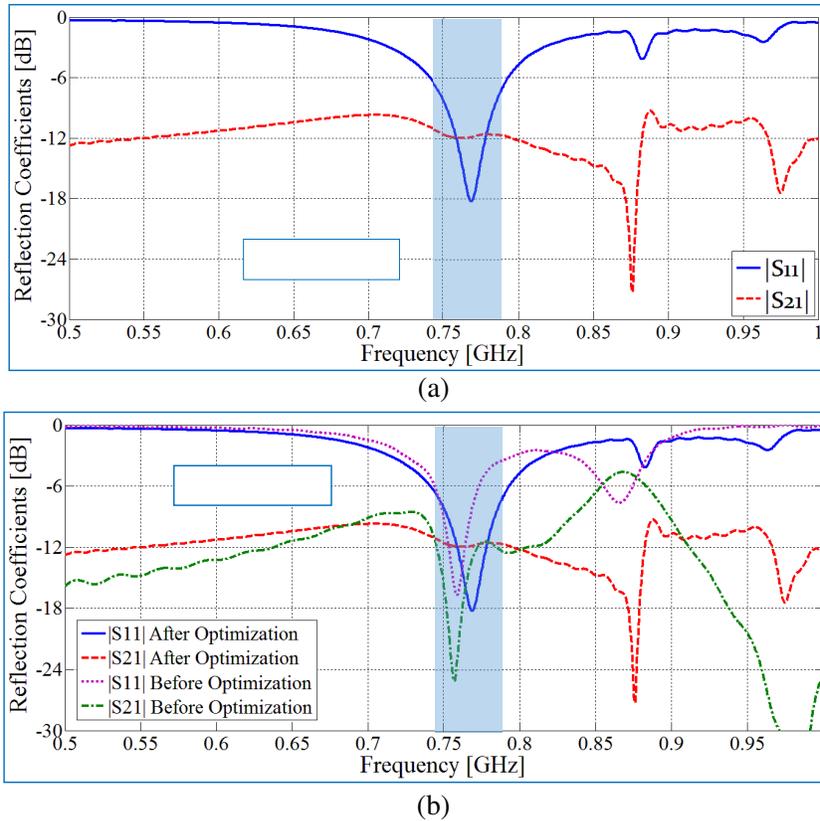


Figure 11. Reflection and coupling coefficients of the simulated antennas, (a) after optimization and (b) before and after optimization.

The calculated envelope correlation coefficient between the antennas is less than (0.15) over the band (741–793) MHz and its minimum value is (0.002) at the frequency (773.5) MHz, as shown in Fig. 12 which meets the required specification of ($ECC < 0.5$).

Figure 13 shows the current distribution on the radiating elements and the deployed decoupling structures namely; slots, N-line and the fork-shaped line. It can be seen that when one antenna element is excited the current density on the other one is small confirming the reduced coupling between the two antenna elements of the MIMO arrangements.

The gain and total efficiency were calculated and shown in Fig. 14. The total efficiency shows peaking around the frequency of 770 MHz and its value is about (65% to 85%) over the band (741–793) MHz while the gain is about (1.88–1.94) dB over the same band.

Figure 15 shows details of the radiation pattern of antenna 1. The other antenna (antenna 2) shows similar patterns. Fig. 15(a) illustrates the 3D variation of the E_ϕ component of the electric field. The pattern has a broad beam with a maximum in the perpendicular direction to the antenna plane. The E_ϕ field component vanishes along a narrow angle in the direction along the YZ -Plane. The insert plot shows a -30 dB null along the Y -direction. Fig. 15(b) shows the E_θ component of the electric field. The pattern is omni-directional in the plane normal to the antenna. The insert plot shows that there is only 1.25 dB variation of the field in the XY -Plane. This is a desired property for mobile handset applications. It can be noticed that the maximum value of the E_ϕ component is larger than the maximum value of the E_θ component by 2.7 dB indicating a relatively high cross polarized component.

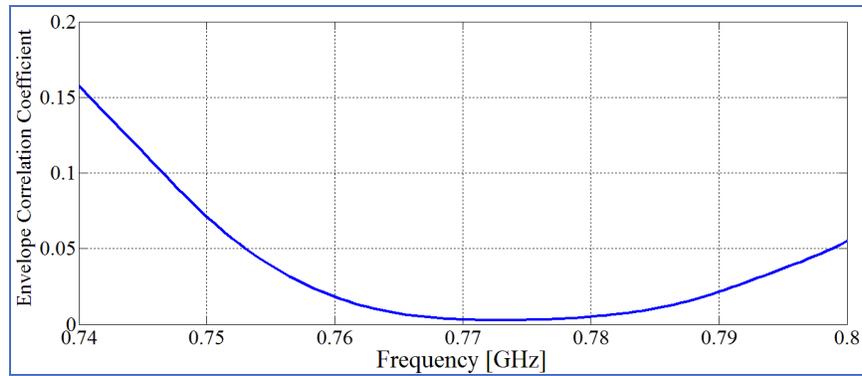


Figure 12. Correlation coefficient of antennas in Fig. 9(b).

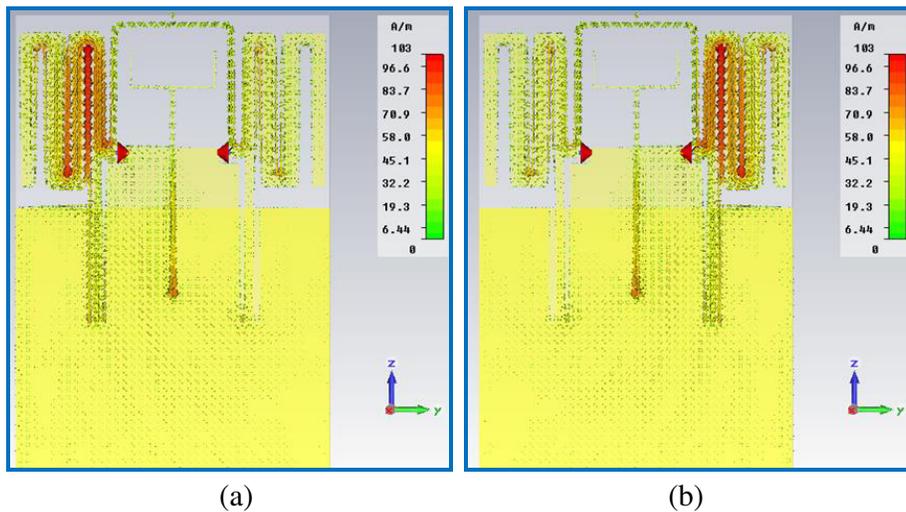


Figure 13. Surface current distribution at 770 MHz when (a) antenna 1 was excited while antenna 2 was match-loaded and (b) antenna 2 was excited and antenna 1 was match-loaded.

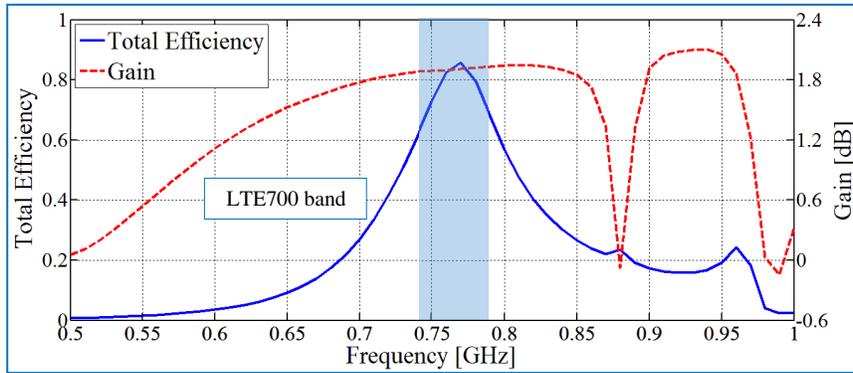
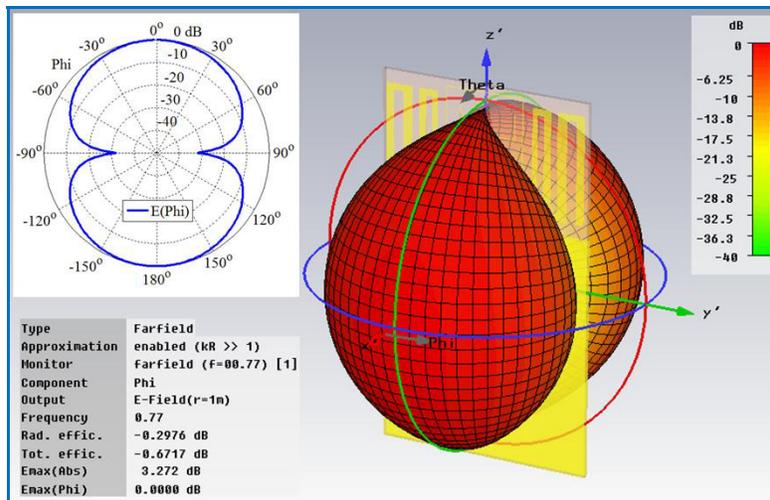
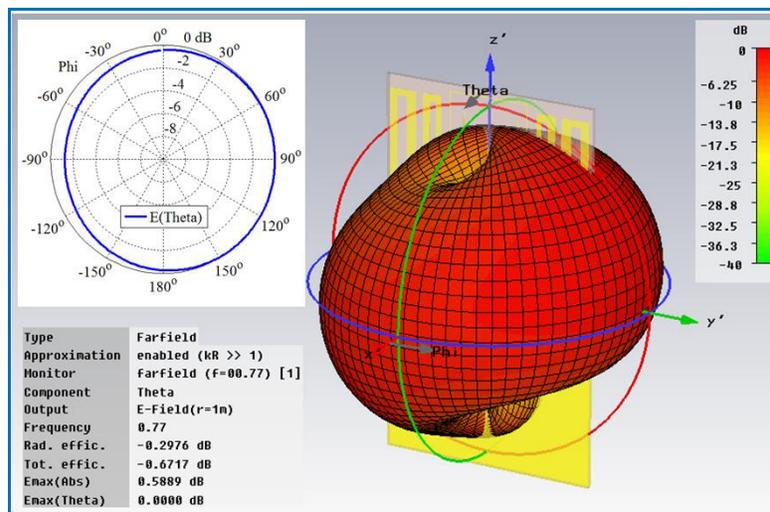


Figure 14. The gain and the total efficiency of antennas in Fig. 9(b).



(a)



(b)

Figure 15. The simulated radiation patterns of the meandered-line PIFA antenna. The insert shows a planar display in the XY-Plane. (a) Variation of the E_ϕ component of the electric field. (b) Variation of the E_θ component of the electric field.

5. EXPERIMENTAL RESULTS OF MIMO ANTENNAS WITH DECOUPLING ARRANGEMENTS

The designed MIMO antennas with the decoupling arrangements were fabricated and tested at the Industrial Research and Development Administration at the Ministry of Science and Technology/Baghdad, see Fig. 16. The antenna was fabricated using an FR4 substrate of relative permittivity 4.3. The overall dimensions of the MIMO antenna system is $110 \times 65 \times 1.6 \text{ mm}^3$.

The performance of the MIMO antenna was measured using a vector network analyzer and the obtained results are compared with the simulation results as shown in Fig. 17. The figure shows that the -6 dB bandwidth of the simulation results is (52 MHz) for the band (741–793) MHz while for the measured results, it is (75 MHz) for the band (720–795) MHz. The maximum simulated mutual coupling at this band is (-11 dB), while the measured mutual coupling at this band is slightly better at the level of (-12.5 dB).

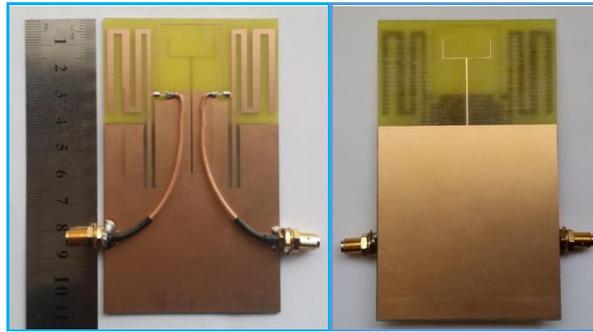


Figure 16. Prototype of the two meandered line PIFA antennas.

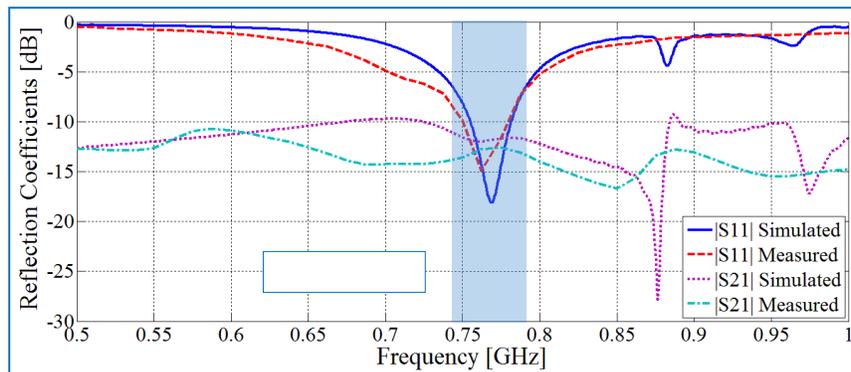


Figure 17. The simulated and the measured reflection and coupling coefficients of the MIMO antennas of Fig. 9.

6. COMPARISON WITH PUBLISHED WORK

The obtained results of the fabricated meandered-line PIFA antenna system is compared with those working at the same LTE band (band class 13) that were published in [12, 13, and 16–23]. The comparison takes into consideration the total substrate dimensions and the volume of the handset, frequency range, bandwidth, maximum coupling between antennas, envelope correlation coefficient, gain and radiation efficiency, as shown in Table 3. From Table 3, it can be noticed that the volume of the proposed MIMO antenna system is slightly larger than the MIMO antennas reported in [18, 21–23], but smaller than the volume of the six other ones.

Table 3. Performance comparison of the proposed MIMO meandered-line PIFA antenna with other related works.

Ref.	Dimensions [mm]	Handset volume [mm ³]	Frequency range [MHz]	-6 dB BW [MHz]	S_{21} [dB]	ECC	Gain [dBi]	Effic. %
[12]	55 × 110 × 5	30250	740–791	51	-31	< 0.4	-2.45	35
[13]	50 × 100 × 3.2	16000	712–752	40	-20	NA	-5.31	NA
[16]	40 × 80 × 5	16000	735–805	70	-15	0.425	-3.59	NA
[17]	55 × 110 × 3	18150	682–787	105	-16.4	0.02	-8.83	NA
[18]	55 × 100 × 1.6	8800	745–796	51	-13	0.05	-4	20
[19]	60 × 100 × 5	30000	738–788	50	-24.6	< 0.1	NA	45
[20]	55 × 88 × 5	24200	735–790	55	-12	< 0.4	-4.7	NA
[21]	60 × 110 × 1	6600	746–787	41	-15	< 0.35	-4.3	NA
[22]	70 × 110 × 0.8	6160	704–787	-5 dB BW 83	-15	< 0.25	NA	30–40
[23]	60 × 120 × 0.8	5760	747–787	40	< -10	0.45	-6.8	40–50
This work	65 × 110 × 1.6	11440	720–795	75	< -12.5	< 0.16	1.94	85

The bandwidth of the proposed MIMO Meandered-Line PIFA antenna is much better than most of the reported MIMO antennas. The proposed MIMO antenna has coupling value that is slightly better than those reported in [20 and 23] while the envelope correlation coefficient is better than those reported in [12, 16, 20–22 and 23]. The gain and the radiated efficiency are the best among all the ten reported works that are listed in Table 3.

7. CONCLUSIONS

An MIMO Meandered-Line PIFA antenna with decoupling arrangements for LTE-700 band has been presented. The achieved bandwidth of the proposed antenna has covered the frequency range (741–793 MHz). The isolation between the two antennas was enhanced using the proposed decoupling arrangements by about 6 dB. The MIMO Meandered-Line PIFA antenna was fabricated using an FR4 substrate of relative permittivity 4.3. Good agreement was found between the measured and simulated results. The antenna has 1.94 dB gain, total efficiency of 85%, and volume of $110 \times 65 \times 1.6 \text{ mm}^3$ that is $(0.275 \times 0.1625 \times 0.004)$ in wavelengths. The proposed MIMO antenna system is suitable for commercial mobile phone applications.

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