

AN ELABORATE FREQUENCY OFFSET ESTIMATION FOR OFDM SYSTEMS

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Abstract—In this paper, a carrier frequency offset estimation scheme in orthogonal frequency division multiplexing is proposed. We focus on increasing the correlation of each PN sequence code repeated in one preamble. We aim at getting an efficiency like using many preambles. The proposed method can improve the performance of the system by estimating fine frequency offset. Also, the proposed method enhances reliability by maximizing the number of correlations compared with the established method in time domain.

1. INTRODUCTION

The orthogonal frequency division multiplexing (OFDM) system, which is now considered as an efficient method for high rate data communication, has some advantages, for example, flexibility which is obtained against the frequency selective channel, high spectrum efficiency [1, 2], and large system capacity [3–8]. So, OFDM system was selected as the standards of IEEE 802.11a, IEEE 802.11g, high wireless LAN of HIPERLAN/2, IEEE 802.16 BWA, DAB and DVB-T which need high data rate [9–12]. But, main disadvantage is that OFDM system is sensitive against frequency offset [13]. In general, time offset in OFDM system is less sensitive than single carrier system [14]. If the initial point of symbol is not correct, it causes decrease of system performance by intercarrier interference (ICI) and intersymbol interference (ISI) [15]. The frequency offset is caused by Doppler shift and carrier frequency offset [16]. Interference damages orthogonality among the sub-carriers. The frequency offset estimation method uses a cyclic prefix (CP) or a preamble. Many methods [17–19] use CP. The frequency offset estimation using the CP is included in data packet. A

preamble can be also used for frequency offset estimation [20–23]. A preamble is used for estimating the offset of one packet itself. We can use moose Method which uses more than two preambles. In this paper, we propose that the pseudo noise (PN) sequence is repeated four times to one packet like [20]. The conventional frequency offset estimation method is compared with the proposed method.

This paper is organized as follow: In Section 2, OFDM system model and frequency offset are introduced. The proposed method is explained in Section 3. In Section 4, we explain the performance evaluation between the conventional method and the proposed method. In Section 5, we conclude about the proposed skill.

2. SYSTEM MODEL DESCRIPTION

We consider an OFDM system implemented by the inverse fast Fourier transform (IFFT) and fast Fourier transform (FFT) [24]. The N symbols are modulated where N is the number of sub-carriers. The transmitted OFDM signal through channel is defined as

$$x(t) = \sum_{l=-\infty}^{\infty} \sum_{k=0}^{N-1} X_l(k) \exp(j2\pi f_k(t - lT_s)) \quad (-T_G \leq t \leq T), \quad (1)$$

where $X_l(k)$ at the k -th sub-carrier in the l -th symbol, means the transmitted data symbol. f_k means k -th sub-carrier. T_s and T_G mean a period of the OFDM symbol the and length of guard interval (GI) respectively. The OFDM symbol is affected by multi-path channel. For prevention of interference between OFDM symbols, GI is interpolated between continuous packets. At this moment, the length of GI is longer than maximum delay propagation of wireless channel. ICI generated by ISI and the delay of sub-carrier can be decreased if CP can be interpolated to GI. The OFDM symbol, which is added by CP in the period of first symbol is changed to discrete signal for processing the continuous signal as follows

$$x_l(n) = \sum_{k=0}^{N-1} X_l(k) e^{j2\pi kn/N} \quad (-N_G < n < \mathbf{N}), \quad (2)$$

where N and N_G are the numbers of samples about valid data of the OFDM symbol and CP respectively. In Eqn. (2), $x_l(n)$ excepting GI becomes IFFT of $X_l(k)$. Therefore, $x_l(n)$ is realized by IFFT of $X_l(k)$. When the frequency offset is estimated ideally, the received signal is expressed as follows:

$$y_l(n) = x_l(n) * h_l(n) + w_l(n), \quad (3)$$

where channel impulse response with multi-path is expressed by $h_l(n)$, and additive white Gaussian noise (AWGN) is also expressed by $w_l(n)$. $y_l(n)$ is can be expressed as

$$y_l(n) = \sum_{k=0}^{N-1} X_l(k)H_l(k)e^{j2\pi kn/N} + w_l(n), \quad -N_G \leq n \leq N. \quad (4)$$

The OFDM system is very sensitive to the frequency offset. The orthogonality is destroyed when frequency offset raises in the OFDM system. The frequency offset is generated by mismatching of exact synchronization between transmitter and receiver. We define that fine frequency offset is called δ and the signal without noise is transmitted, $Y_l(k)$ expressed in frequency domain as follows

$$Y_l(k) = X_l(k)H_l(k) \frac{\sin \pi \delta}{N \sin \frac{\pi \delta}{N}} e^{j\pi \delta (N-1)/N} + I_l(k), \quad (5)$$

where ICI is expressed by $I_l(k)$ that is generated by fine frequency offset [25].

3. FREQUENCY OFFSET ESTIMATION

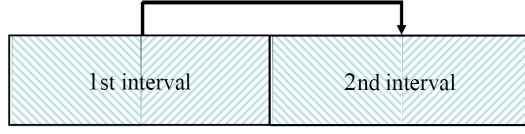
A preamble is used for frequency offset estimation. The preamble using Schmidl estimation method is expressed as a reiterated form in time domain. The fine frequency offset can be estimated by the correlation value that is obtained through the operation between first interval and second interval in Schmidl structure. The process can be written as

$$\hat{\delta}_{schmidl} = 2 \times \frac{1}{2\pi} \arg \left(\sum_{n=0}^{\frac{N}{2}-1} r^*(n)r \left(n + \frac{N}{2} \right) \right). \quad (6)$$

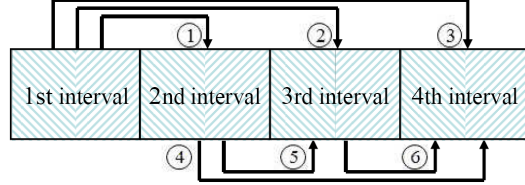
Schmidl structure is depicted to Fig. 1(a). The preamble using proposed estimation method has the form which is reiterated with 4 times using PN sequence excepting CP interval in time domain. The proposed method is depicted to Fig. 1(b). This preamble to adapt proposed method consists of CP, first interval, second interval, third interval and fourth interval.

How to estimate fine frequency offset is the method of using the correlation between each PN sequence in the preamble. Let us consider step-by-step process,

step 1) The part of the frequency offset is estimated by the correlation that is calculated between first interval and the others(second/third/fourth intervals).



(a) Schmidl method



(b) Proposed method

Figure 1. Comparison of structure of frequency offset estimation method.

step 2) The second interval correlate with third/fourth intervals for estimating frequency offset.

step 3) The third interval is calculated to fourth interval.

step 4) Each value from step 1 to step 3 is averaged.

Through above procedure, we can estimate the accurate frequency offset as

$$\hat{\delta}_{proposed} = \frac{1}{M} \sum_{s=1}^{L-1} \sum_{l=s}^{L-1} \left(\frac{L}{l} \frac{1}{2\pi} \times \arg \left(\sum_{n=0}^{\frac{N}{L}-1} r^* \left(n + (s-1) \frac{N}{L} \right) \right) \times r \left(n + (s-1) \frac{N}{L} + \frac{N}{L} \right) \right), \quad (7)$$

$$M = \sum_{l=1}^{L-1} l, \quad (8)$$

where s is the index for calculation, L is the number of intervals in a preamble and M is the number of operation of the correlation between each interval.

4. SIMULATION RESULTS AND DISCUSSIONS

In this section, simulation was performed with the following parameters: the number of sub-carriers $N = 64$, frequency offset

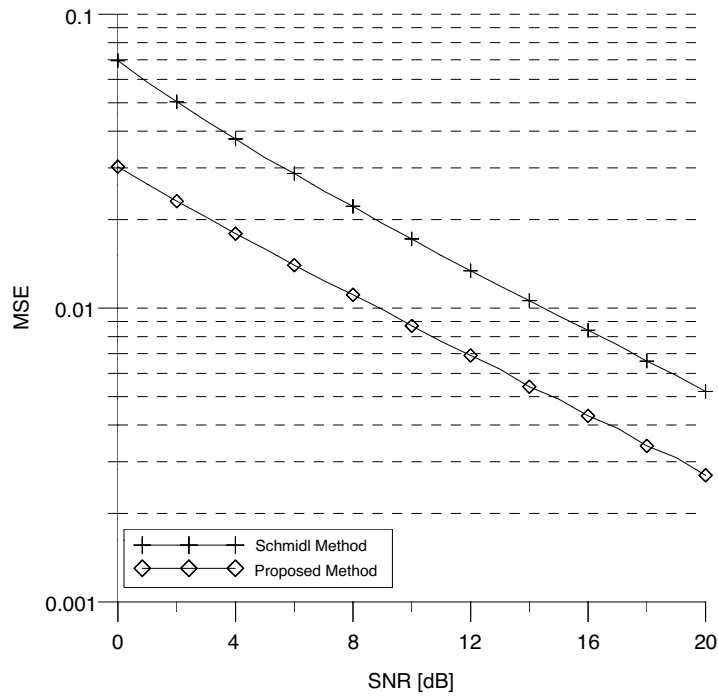


Figure 2. The estimation variance in AWGN and one path channel.

$f_0 = 0.01$. The channel is applied to the simulation which does not existed attenuation, also we assume that the channel estimation and time synchronization is perfect. When the proposed method is applied, the BER performance is more similar to perfect estimation than using Schmidl method. The mean square error (MSE) of estimated frequency offset is shown in Fig. 2. The fine frequency offset estimation accuracy of Schmidl method and proposed method in transmit power 0 dB is shown in Fig. 3. The bit error rate (BER) of data packet which is recovered by estimated frequency offset is shown in Fig. 4. The proposed scheme's SNR gain of 0.4 dB is archived in the BER 10^{-2} . In the proposed method, the more the repeated interval increases, the more it shows similar BER curve to the recovery by perfect estimation. So, when we increase the number of repeated interval and enlarge the computation from the correlation, it can be guessed that it is similar to the recovery by fine frequency offset. When we estimate the fine frequency offset, we can find out that there is a tradeoff between complexity and performance.

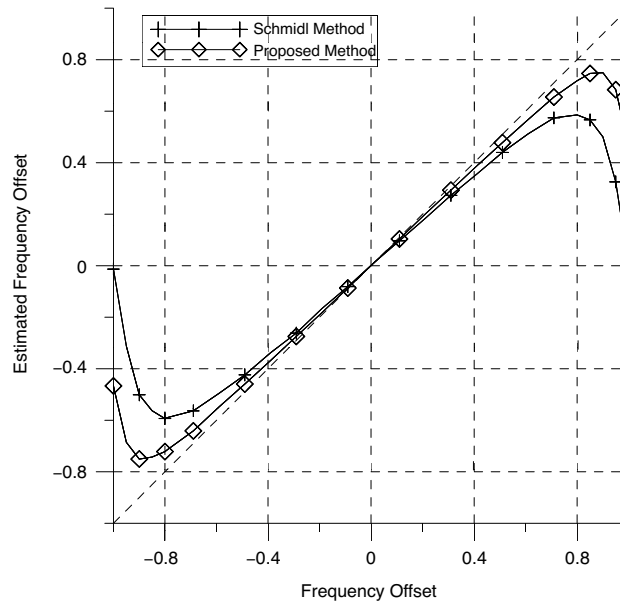


Figure 3. Comparison of estimation performance for several methods at SNR = 0 dB.

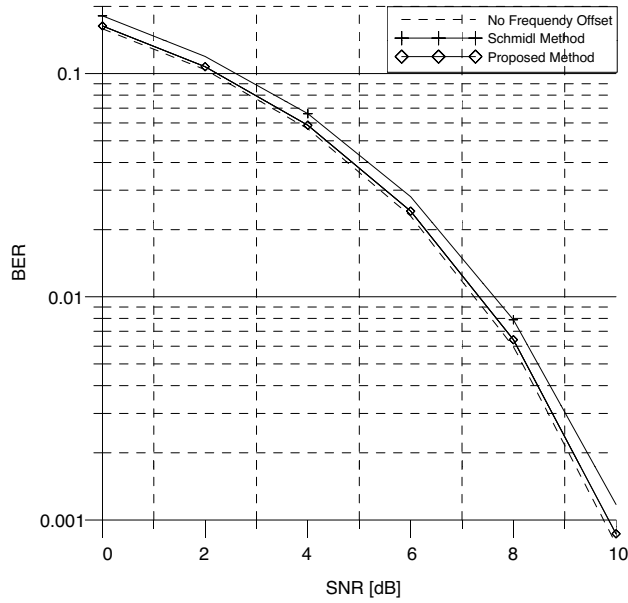


Figure 4. The estimation BER in AWGN and one path channel.

5. CONCLUSIONS

In this paper, a new method of frequency offset estimation is proposed. The proposed method provides more accurate estimation performance than conventional method. In the future, using the same structure, we adapt to coarse frequency offset estimation. Moreover, we will study the method reducing complexity to get accurate frequency offset estimation.

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