

## **STUDY ON THE IMPEDANCE-MATCHING TECHNIQUE FOR HIGH-TEMPERATURE SUPERCONDUCTING MICROSTRIP ANTENNAS**

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**Abstract**—Impedance-Matching technique is in common use for antennas to broaden their bandwidth. Its application in high-temperature superconducting microstrip antennas is studied theoretically in this paper. It is found that employing an impedance-matching network directly to HTS microstrip antennas to broaden their bandwidth is of little significance.

### **1. INTRODUCTION**

Due to the much lower surface resistance compared with normal metals and high transition temperature over the boiling point of liquid nitrogen, the high-temperature superconductor (HTS) has found wide applications in microwave circuits and antennas [1]. A high-temperature superconducting microstrip antenna (HTSMA) can obtain a rather high gain, but suffers from a very narrow bandwidth, which severely limits its application. Generally, the bandwidth of HTSMAs is only 0.85%–1.1% [2], which is even narrower than that of its normal counterparts.

There are many techniques [3–7] to enhance the bandwidth of conventional microstrip antennas, among which the impedance-matching technique [3] is also commonly used. It has been shown that some of these methods are also efficient for HTSMAs [8].

In this paper, the theory associated with impedance-matching techniques is proposed and the feasibility of employing the technique in HTSMAs is investigated theoretically.

## 2. THEORY ASSOCIATED WITH THE IMPEDANCE-MATCHING TECHNIQUE

In the vicinity of its fundamental resonant frequency, the input impedance of a microstrip antenna can be modeled by a parallel-resonant RLC and expressed by [3]

$$Z_{in} = \frac{R_0}{1 + jQv} \quad (1)$$

where  $R_0$  is the resonant resistance,  $Q$  is the quality factor and

$$v = \frac{f}{f_r} - \frac{f_r}{f} \quad (2)$$

Here  $f$  is the frequency variable and  $f_r$  is the resonant frequency. If the feed line has a characteristic impedance  $Z_0$ , the input VSWR  $\rho$  is given by

$$\left| \frac{Z_{in}(f) - Z_0}{Z_{in}(f) + Z_0} \right| = \frac{\rho(f) - 1}{\rho(f) + 1} \quad (3)$$

If the bandwidth criterion is taken to be  $\rho \leq S$ , and  $f_1$  and  $f_2$  are the lower and upper band edge frequencies respectively, so that  $\rho(f_1) = \rho(f_2) = S$ , the relative bandwidth is given by

$$B = \frac{f_2 - f_1}{f_r} \quad (4)$$

It can be derived that

$$B = \frac{1}{Q} \sqrt{\frac{(TS - 1)(S - T)}{S}} \quad (5)$$

where  $T = R_0/Z_0$ . It follows from (5) that to maximize  $B$ ,  $T$  must be equal to its optimal value, which is given by

$$T_{opt} = \frac{1}{2} \left( S + \frac{1}{S} \right) \quad (6)$$

It is evident that the bandwidth given by (5) can be increased, at least in principle, by using an impedance matching network. The best

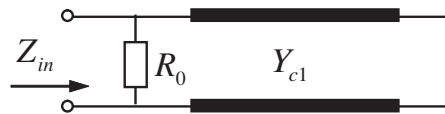
situation of utilizing an impedance-matching network is to realize a constant (but not perfect) match within the band of operation and a total mismatch outside the band. The maximum  $\rho \leq S$  bandwidth obtainable for a parallel-resonant circuit is given by [3]

$$B_m = \frac{1}{Q} \frac{\pi}{\ln[(S+1)/(S-1)]} \quad (7)$$

For a microstrip antenna, the quality factor  $Q$  can be expressed by

$$Q = \frac{\pi R_0}{2 Z_{c1}} \quad (8)$$

where  $Z_{c1} = 1/Y_{c1}$  is the equivalent characteristic impedance of the microstrip antenna in the transmission line model (shown in Figure 1).



**Figure 1.** The transmission line model for microstrip antennas.

### 3. FEASIBILITY INVESTIGATION OF THE IMPEDANCE-MATCHING TECHNIQUE FOR HTSMAS

Usually a HTS substrate has a rather high permittivity which will lead to a considerably large resonant resistance for HTS microstrip antennas. According to the theory proposed above, the maximum bandwidth will be affected.

A rectangular patch antenna with a typical HTS material is analyzed. The length and width of the antenna is  $935 \mu\text{m}$  and  $1630 \mu\text{m}$  respectively. The typical HTS film has a critical temperature of 89 K and magnetic penetration depth of 140 nm at 0 K.

The resonant resistance of the antenna, which is estimated from a measured result at 77 K, is  $1690 \Omega$  [2]. Using a transmission line model [9] the equivalent characteristic impedance of the HTS microstrip antenna is calculated and shows a value of  $9.4 \Omega$ . Taking these values into Equation (8) and Equation (7), we find that the maximum bandwidth of  $\rho \leq 2$  is only 0.01.

Since the structure and the HTS material of the antenna here are very typical, it indicates that employing impedance-matching networks directly to a HTS microstrip antenna to broaden its bandwidth is not so efficient.

#### 4. CONCLUSION

To investigate the feasibility of employing impedance-matching techniques to broaden the bandwidth of a HTS microstrip antenna, a typical HTS microstrip antenna is taken as an example. The numerical result shows that utilizing an impedance-matching network directly to HTS microstrip antennas is not an efficient method to improve the frequency bandwidth.

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