Surface Mounted Microstrip Antenna Using Ball Grid Array Packaging for mmWave Systems Integration

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Abstract—In this letter, two cost-effective surface-mount patch antenna elements for millimeter-wave (mmWave) systems using ball grid array (BGA) packaging are presented. A single-layer substrate based on FR4 is used to meet the low-cost requirements. The BGA packaging makes the proposed antenna element compact and easy to integrate. A U-slot is added to the patch to improve the impedance bandwidth of the patch antenna, and a vertical transition is designed to transmit the excitation signal by using a plated through-hole (PTH). The design process of the antenna is illustrated in detail. The antenna prototype has been simulated, fabricated, and measured to validate the design. The size of the fabricated prototype is $5 \text{ mm} \times 5 \text{ mm} \times 1.3 \text{ mm}$, which is very suitable for integration into a mmWave system.

1. INTRODUCTION

Microstrip antennas are the most widely used antennas in wireless systems because of their advantages such as low cost, small size, and convenient manufacture [1-3]. To increase the bandwidth, a U-slot is added to improve the bandwidth of the patch antenna [4-7]. However, the main disadvantage of the above-mentioned patch antenna is the use of high-loss coaxial cables for interconnection. On the one hand, the coaxial cable increases the insertion loss between the antenna and RF amplifier. On the other hand, the large size is difficult to integrate into the system.

Antenna-in-package (AiP) is a highly integrated solution that can achieve the low-loss interconnection between the antennas and RF chipsets [8–10]. In addition, flip-chip interconnect can reduce insertion loss and improve RF system performance, instead of using bulky and lossy coaxial connectors [11] or waveguide connectors [12].

In this letter, two small-size and low-cost BGA packaged patch antenna elements are proposed. Taking advantage of the advantages of the patch antenna and AiP technology, the compact BGA packaged antenna element is suitable for integration with other surface mount devices (SMDs) in the same package. The simulated results and experimental results are given and discussed. The antenna elements have been designed, fabricated, and measured. The antenna has a wide bandwidth and a stable radiation pattern. All simulation results are given by ANSYS electromagnetics.

2. ANTENNA GEOMETRY AND DESIGN

Figures 1(a), (b), and (c) show the geometry of the proposed patch antenna element without the U-slot (Ant 1). Figures 1(d), (e), and (f) show the geometry of the proposed patch antenna element with the U-slot (Ant 2). The dimensions of the antenna elements are the same, and Ant 2 adds a U-slot based on Ant 1. The antenna element is fabricated on a low-cost FR4 substrate with a dielectric constant

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Figure 1. Geometry of the proposed antenna element. Ant 1: (a) Explored view. (b) Top view. (c) Bottom view. Ant 2: (d) Explored view. (e) Top view. (f) Bottom view.

of 4.25, dielectric loss tangent of 0.018, and thickness of 1.1 mm. Two conductive layers are realized on the top and bottom layers of the antenna element. To obtain a directional radiation pattern, the bottom metal acts as a reflective ground. The 0.2 mm plated through-hole (PTH) is a vertical transition that connects the patch and the feedline at the bottom layer. The length of the feedline is a quarterwavelength and is used for impedance matching. The solder resists layer is coated on the bottom metal layer. Solder balls with a diameter of 300 microns are mounted on the bottom layer of the element. In addition, to facilitate integration with the RF chipset, the input impedance of the patch element is set to 50Ω . After electromagnetic simulation, the dimensions are summarized in Table 1. The final size is $5 \text{ mm} \times 5 \text{ mm} \times 1.3 \text{ mm}$.

Table	1.	Dimensions	of	the	proposed	$\operatorname{antenna}$	element (Units:	mm)
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Parameters	Values	Parameters	Values
L1	5	W1	5
L2	2.2	W2	0.8
L3	1.3	W3	0.2
L4	1.5	W4	2.8

Figure 2 shows the surface current distribution of the antenna element at 32 GHz. It indicates that the patch is strongly excited by the PTH. The U-slot adds a path and enhances the current intensity. Figure 3 shows the performance of the prototype with and without the U-slot. It can be seen that the -10 dB impedance bandwidth of Ant 1 ranges from 30.9 to 37.4 GHz. Due to the addition of U-slots, the bandwidth of Ant 2 is wider than that of Ant 1. The -10 dB impedance bandwidths of Ant 2 are 27.7 to 32.5 GHz and 34 to 36.9 GHz. In addition, Ant 1 has a gain range of 4.11 to 6.43 dBi, and Ant 2 has a gain range of 3.7 to 6.88 dBi. In the range of 28 to 38 GHz, the simulation efficiencies of Ant 1 and Ant 2 reach more than 71.9% and 59.2%, respectively.

3. MEASUREMENT RESULTS AND DISCUSSION

To verify the proposed design, the prototype was fabricated using the low-cost printed circuit board (PCB) technique, as shown in Figure 4. An evaluation board with a 50 Ω coplanar waveguide (CPW) transmission line was used to measure the proposed antenna element. Furthermore, the proposed



Figure 2. Surface current distribution of the proposed antenna element at 32 GHz. (a) Ant 1. (b) Ant 2.



Figure 3. (a) Simulated reflection coefficient of the proposed antenna. (b) Simulated peak gain and efficiency of the proposed antenna.



Figure 4. Photograph of the patch antenna prototype. Ant 1: (a) Antenna element. (b) Assembly prototype. Ant 2: (c) Antenna element. (d) Assembly prototype.

antenna element is surface-mount on the top layer of the evaluation board after reflow-soldering. In addition, a 2.92 mm end launch connector with operating frequencies from DC to 40 GHz was installed at the edge of the evaluation board. The reflection coefficient is measured by Rohde & Schwarz Network Analyzer (ZVA40). The far-field radiation patterns are measured in an anechoic chamber.

Figures 5(a) and (c) show the simulated and measured reflection coefficients and gains of Ant 1, respectively. The measured impedance bandwidth of -9.72 dB is in the range of 30.8 to 35.7 GHz, which is slightly narrower than the simulated bandwidth. In the range of 28 to 38 GHz, the measured peak gain is 3.18 to 4.22 dBi. Figures 5(b) and (d) show the simulated and measured reflection coefficients and gains of Ant 2, respectively. The measured impedance bandwidth of -10 dB ranges from 30.9 to 36 GHz. The measured gain ranges from 3.19 to 4.97 dBi. Figure 6 shows the simulated and measured *E*-plane and *H*-plane normalized radiation patterns at 32 GHz, 34 GHz, and 36 GHz, respectively.

It can be seen that there are discrepancies between the simulated and measured results. As mentioned in [1], the tolerance control has a great influence on the performance of the microstrip antenna in the millimeter frequencies. Through the analysis of the parameters, we find that the difference is



Figure 5. (a) Measured and simulated reflection coefficient of the Ant 1. (b) Measured and simulated reflection coefficient of the Ant 2. (c) Measured and simulated peak gain of the Ant 1. (d) Measured and simulated peak gain of the Ant 2.

Ref.	Antenna type	Fc	Imp. BW (-10 dB) (%)	Measured peak gain (dBi)	$\begin{array}{c} \text{Dimensions} \\ (\text{mm}^3) \end{array}$	Material	Interconnection
[11]	Huygens source 27.9		2.14	4.54	$2.4 \times 2 \times 1.143$	Rogers5880	Coaxial line
[12]	Patch	27.9	8.6	7.41	$19.9\times 30\times 0.79$	TLY-5	Waveguide
This work	Patch (Ant 1)	33.35	14.7 *	4.22	$5 \times 5 \times 1.3$	FR4	Surface-mount
	Patch (Ant 2)	33.45	15.2	4.97	$5 \times 5 \times 1.3$	FR4	Surface-mount

Table 2. Comparisons between the proposed and reported antennas.

* _	9.	72	dB
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mainly caused by the fabrication error, such as the thickness of the substrate and the position of the PTH. The radiation pattern of the E-plane is slightly tilted due to the 2.92 mm connector.

Table 2 summarizes the performance comparison between the recommended antenna and the reported antenna. It can be seen that the antennas in [11] and [12] have to use bulky and lossy waveguide connectors or coaxial cables to interconnect with the RF chipset. Meanwhile, the proposed antenna is a BGA packaged element that can be surface-mounted with other active devices or integrated into the system without using connectors. Low-cost FR4 substrate is also an attractive material to meet cost-effectiveness requirements.





Figure 6. Simulated and Measured *E*-plane and *H*-plane normalized radiation patterns at 26 GHz, 28 GHz and 30 GHz.

4. CONCLUSIONS

A surface-mount microstrip antenna based on the BGA packaging has been proposed. The prototype has been simulated, fabricated, and measured. Based on the BGA packaging, the antenna element achieves a compact size and surface-mount function. Cost-effective functions can be obtained by using a single-layer low-cost FR4 printed circuit board. In addition, the performance of the patch antenna will be sensitively affected by mechanical manufacturing errors. Therefore, to obtain good performance, manufacturing tolerances should be strictly limited. The proposed antenna is a good choice for integration into millimeter-wave systems.

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Progress In Electromagnetics Research Letters, Vol. 98, 2021

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